

1) $w = 6 \text{ mm}$, $d = 0.25 \text{ mm}$, $t = 25 \text{ mm}$, $\sigma_c = 5.5 \times 10^7 \text{ S/m}$, $\sigma = 3.5 \times 10^{-3} \text{ S/m}$,
 $\epsilon = 25 \text{ pF/m}$, $f = 750 \text{ MHz}$, $\mu = \mu_c = \mu_0 = 4\pi \times 10^{-7} \text{ H/m}$

$$\text{Resistance } R (\Omega/\text{m}) = \frac{2}{w\delta\sigma_c} = \frac{2\sqrt{\pi f \mu_c \sigma_c}}{w\sigma_c} = \frac{2\sqrt{\pi f \mu_c}}{w\sqrt{\sigma_c}} = 2.445 \Omega/\text{m}$$

(for $\delta \ll t$)

$\delta \rightarrow \text{sk in depth}$

$$\text{Inductance } L (\text{H/m}) = \frac{\mu d}{w}$$

Since, $\delta = \frac{1}{\sqrt{\pi f \mu_c \sigma_c}}$

$$= 2.47 \times 10^{-6} \text{ m}$$

$$\delta \ll t$$

$$\text{Conductance } G (\text{S/m}) = \frac{\sigma w}{d}$$

$$\text{Capacitance } C (\text{F/m}) = \frac{\epsilon w}{d}$$

$$\text{Inductance } L = \frac{\mu d}{w} = \frac{4\pi \times 10^{-7} \times 0.25 \times 10^{-3}}{6 \times 10^{-3}} = 5.23 \times 10^{-8} \text{ H/m}.$$

$$= 52.3 \text{ nH/m}.$$

$$\text{Conductance } G = \frac{\sigma w}{d} = \frac{3.5 \times 10^{-3} \times 6 \times 10^{-3}}{0.25 \times 10^{-3}} = 0.084 \text{ S/m}$$

$$= 84 \text{ mS/metre}.$$

$$\text{Capacitance } C = \frac{\epsilon w}{d} = \frac{25 \times 10^{-12} \times 6 \times 10^{-3}}{0.25 \times 10^{-3}} = 600 \times 10^{-12} \text{ F/m}$$

$$= 600 \text{ pF/m}.$$

2a) Propagation constant $\gamma = \sqrt{(R + j\omega L)(G + j\omega C)}$

$$= \sqrt{(2.445 + j \times 2\pi \times 750 \times 10^6 \times 5.23 \times 10^{-8})}$$

$$(84 \times 10^{-3} + j 2\pi \times 750 \times 10^6 \times 600 \times 10^{-12})$$

$$= \sqrt{(2.445 + j 246.457)(84 \times 10^{-3} + j 2.827)}$$

$$\gamma = (697.075 \angle 177.729^\circ)^{1/2}$$

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$$= \sqrt{697.075} \angle \left(\frac{177.729^\circ}{2} \right)$$

$$= 26.4 \angle 88.86^\circ = 0.5252 + j 26.314 \text{ 1/m.}$$

2(b) The characteristic impedance (Z_0) = $\sqrt{\frac{R + j\omega L}{G + j\omega C}}$

$$Z_0 = \sqrt{\frac{2.445 + j 246.457}{84 \times 10^{-3} + j 2.827}}$$

$$= \sqrt{(87.145 \angle 1.133^\circ)}$$

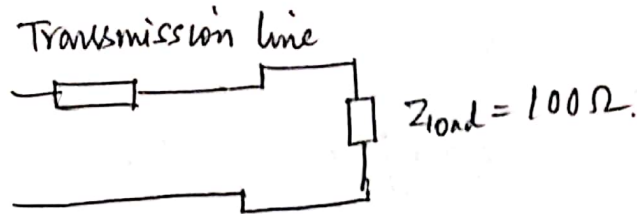
$$= \sqrt{87.145} \angle \frac{1.133^\circ}{2}$$

$$= 9.33 \angle 0.5665^\circ = 9.3295 + j 0.09224 \Omega$$

2(c) When ^{real} power loss in lines is neglected, line resistance can be considered as zero ($R=0$).

When dielectrics are considered to be ideal i.e., they act as a pure capacitance and no real power losses happen ~~due to eddy current or due to~~ due to flow of current through them, conductance is taken as zero ($G=0$).

3. >



Voltage reflection coefficient (Γ_L) when transmission line of characteristic impedance (Z_0) is terminated with a load of impedance (Z_L) is given as —

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{100 - (9.3295 + j0.09224)}{100 + (9.3295 + j0.09224)}$$

$$= 0.8293 - j1.5448 \times 10^{-3}$$

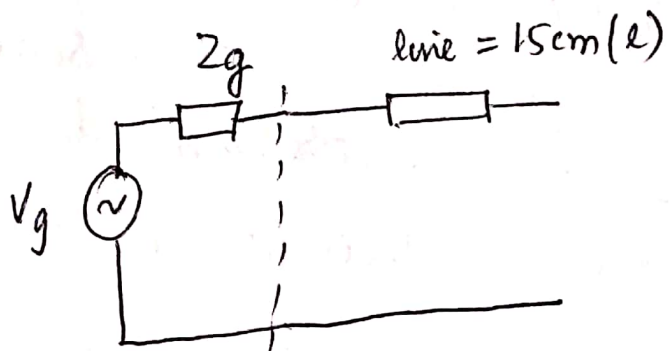
$$= 0.8293 \angle -0.106^\circ$$

Standing wave ratio (~~SWR~~ SWR) = $\frac{1 + |\Gamma_L|}{1 - |\Gamma_L|}$

$$= \frac{1 + 0.8293}{1 - 0.8293}$$

$$= 10.716$$

4. >



$$4.7) Z_{in} = Z_0 \left[\frac{Z_L + Z_0 \tanh(\gamma L)}{Z_0 + Z_L \tanh(\gamma L)} \right]$$

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$$Z_L = 100 \Omega$$

$$Z_0 = (9.395 + j0.09224) \Omega$$

$$\begin{aligned} \tanh \gamma L &= \tanh(\alpha L + j\beta L) \\ &= \frac{\tanh(\alpha L) + j \tan(\beta L)}{1 + j \tanh(\alpha L) \tan(\beta L)} \end{aligned}$$

$$\text{Length of line } (L) = 15 \text{ cm.}$$

$$\therefore \gamma L = (0.5252 + j26.394) \times 0.15$$

$$= 0.078 + j3.9591$$

$$= \alpha L + j\beta L$$

↓ radians

$$\therefore \tanh(\gamma L) = \frac{\tanh(0.078) + j \tan(3.9591)}{1 + j \tanh(0.078) \tan(3.9591)}$$

$$= \frac{0.0778 + j1.0663}{1 + j0.083} = 0.165 + j1.052$$

$$\begin{aligned} Z_{in} &= (9.395 + j0.09224) \left[\frac{100 + (9.395 + j0.09224)(0.165 + j1.052)}{(9.395 + j0.09224) + 100(0.165 + j1.052)} \right] \\ &= \frac{(9.395 + j0.09224)(101.453 + j9.898)}{(25.895 + j105.292)} \end{aligned}$$

$$= (3.0139 - j8.3025) \Omega = 8.83 \angle -1.222 \text{ rad } \Omega$$

4b)

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$$V_0^+ = \left(\frac{Z_m}{Z_m + Z_g} \right) \frac{V_g}{e^{\gamma_L} + \Gamma_L e^{-\gamma_L}}$$

$$e^{\gamma_L} = e^{(0.5252 + j26.394) \times 0.15}$$

$$= e^{0.07878} e^{j3.9591}$$

$$= -0.7401 - j0.789$$

$$e^{-\gamma_L} = e^{(-0.5252 - j26.394) \times 0.15}$$

$$= e^{-0.07878} e^{-j3.9591}$$

$$= -0.6322 + j0.6742$$

$$\Gamma_L = 0.8293 - j1.5448 \times 10^{-3}$$

$$V_0^+ = \frac{(3.0139 - j8.3025) \times 12}{(3.0139 - j8.3025 + 10) \times [(-0.7401 - j0.789) + (0.8293 - j1.5448 \times 10^{-3}) \times (-0.6322 + j0.6742)]}$$

$$= \frac{(3.0139 - j8.3025) \times 12}{(13.0139 - j8.3025) (-1.2633 - j0.228)}$$

$$= -3.5926 + j3.96 = 5.346 \angle 2.307 \text{ rad Volts.}$$

5a)

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$$V_s(z) = V_0^+ e^{-\gamma z} + V_0^- e^{\gamma z}$$

$$\Gamma_L = \frac{V_0^-}{V_0^+}$$

$$= V_0^+ (e^{-\gamma z} + \Gamma_L e^{\gamma z})$$

$$= V_0^+ e^{j\phi_g} \left[e^{-\alpha z - j\beta z} + \cancel{\Gamma_L} \Gamma_L e^{+\alpha z + j\beta z} \right] \quad \phi_g \Rightarrow \text{phase angle of source.}$$

$$\gamma = 0.5252 + j26.394$$

$$= \alpha + j\beta$$

$$\therefore \alpha = 0.5252$$

$$\beta = 26.394$$

$$\phi_g = 0^\circ$$

$$V_0^+ = 5.346 \angle 2.307 \text{ Volts.}$$

$$\Gamma_L = 0.8293 - j1.5448 \times 10^{-3}$$

$$V_s(z) = 5.346 \angle 2.307 \left[e^{(-0.5252 - j26.394)z} + (0.8293 - j1.5448 \times 10^{-3}) e^{(0.5252 + j26.394)z} \right]$$

$$= 5.346 e^{j2.307} \left[e^{(-0.5252 - j26.394)z} + 0.8293 e^{-j0.00186} e^{(0.5252 + j26.394)z} \right]$$

$$= 5.346 e^{(-0.5252 - j26.394)z + j2.307} + 4.433 e^{(0.5252 + j26.394)z - j0.00186}$$

Volts

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$$\begin{aligned} 5b) \quad V(z, t) &= \operatorname{Re}[V_s(z) e^{j\omega t}] \\ &= \operatorname{Re}[V_s(z) e^{j2\pi \times 750 \times 10^6 t}] \\ &= \operatorname{Re}[V_s(z) e^{j4.71 \times 10^9 t}] \end{aligned}$$

$$\begin{aligned} V(z, t) &= 5.346 e^{-0.5252z} \cos(4.71 \times 10^9 t - 26.394z + 2.307) \quad \downarrow \text{radians} \\ &+ 4.433 e^{0.5252z} \cos(4.71 \times 10^9 t + 26.394z - 0.00186) \quad \downarrow \text{radians} \end{aligned}$$

Volts